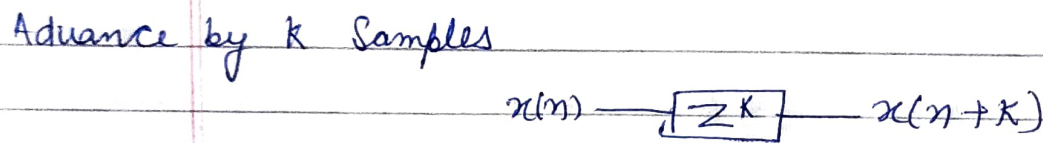
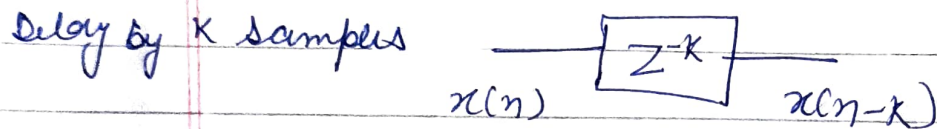
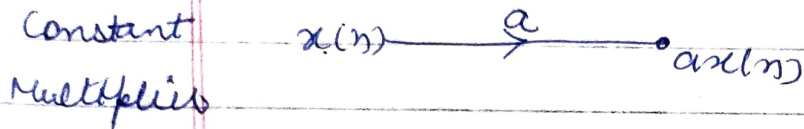
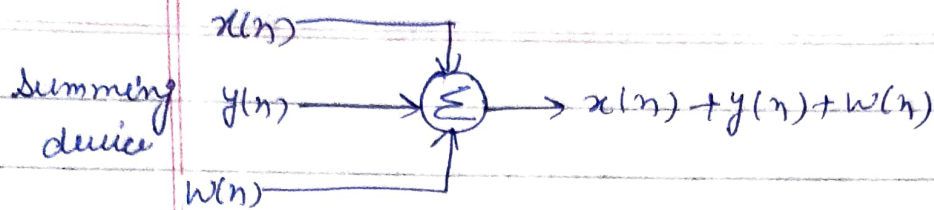
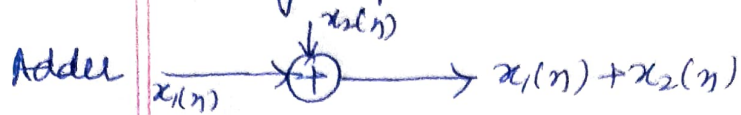


Block Diag Representation



What is the meaning of canonic and non canonic structures $\Rightarrow$

If the number of delay elements in the block diagram is equal to the order of difference equation of a digital filter, then the realization structure is called as canonic structure. Otherwise it is called as non-canonic structure.

FIR Filters

Transfer function of FIR filter →

$$y(n) = \sum_{k=0}^{M-1} h(k) x(n-k) \quad \text{--- (1)}$$

Taking z transform of eqⁿ (1)

$$Y(Z) = \sum_{k=0}^{M-1} h(k) Z^{-k} X(Z) \quad \text{using time shifting property}$$

$$H(Z) = \frac{Y(Z)}{X(Z)}$$

$$H(Z) = \frac{Y(Z)}{X(Z)} = \sum_{k=0}^{M-1} h(k) Z^{-k} \quad \text{--- (2)}$$

Direct form structure →

FIR filter can be obtained by using the eqⁿ of linear convolution.

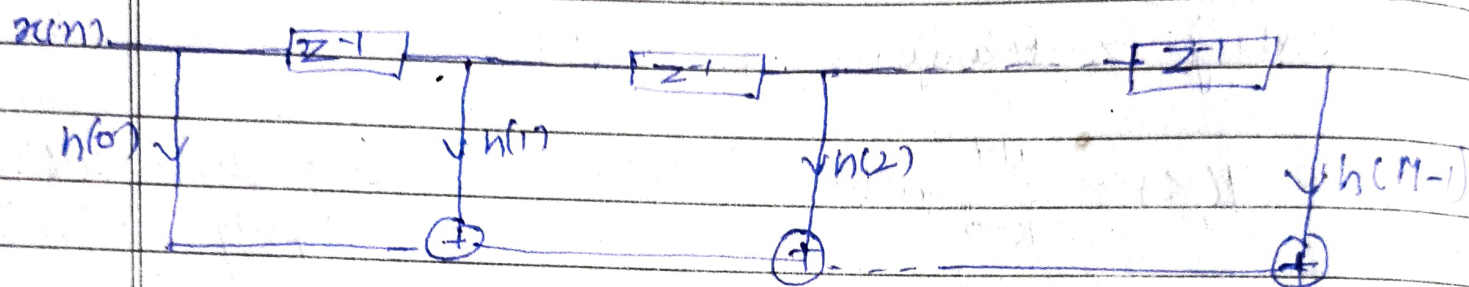
$$y(n) = \sum_{k=0}^L h(k) x(n-k) \quad \text{--- (3)}$$

If we consider that there are M samples in eqⁿ (3)

$$y(m) = \sum_{k=0}^{M-1} h(k) x(n-k) \quad \text{--- (4)}$$

expand eqⁿ (4)

$$y(n) = h(0)x(n) + h(1)x(n-1) + h(2)x(n-2) + \dots + h(M-1)x(n-M+1) \quad (3)$$



Cascade Form Structure $\Rightarrow$

Cascade means the number of stages are connected in series.

$$H(Z) = \sum_{k=0}^{M-1} h(k) Z^{-k} \quad (6)$$

$$H(Z) = H_1(Z) \cdot H_2(Z) \cdot H_3(Z) \dots H_K(Z)$$

$H_K(Z)$ is the second order transfer fn

$$H_K(Z) = h(K_0) + h(K_1)Z^{-1} + h(K_2)Z^{-2} \quad (7)$$

$$H_K(Z) = \frac{Y_K(Z)}{X_K(Z)}$$

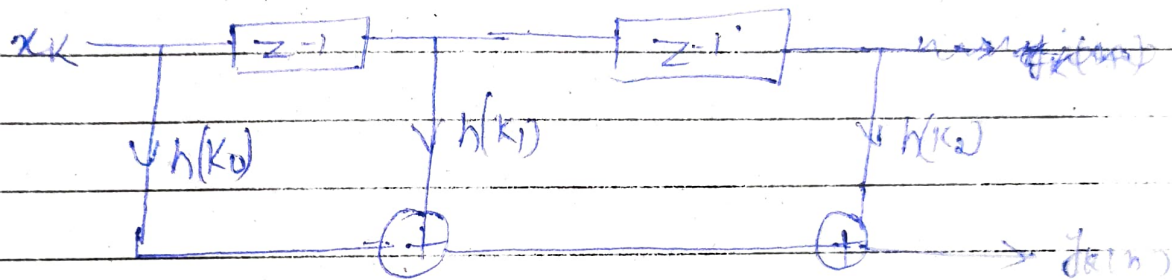
$$\frac{Y_K(Z)}{X_K(Z)} = h(K_0) + h(K_1)Z^{-1} + h(K_2)Z^{-2}$$

$$Y_K(Z) = h(K_0) X_K(Z) + h(K_1) Z^{-1} X_K(Z) + h(K_2) X(Z) Z^{-2} \quad \text{--- (8)}$$

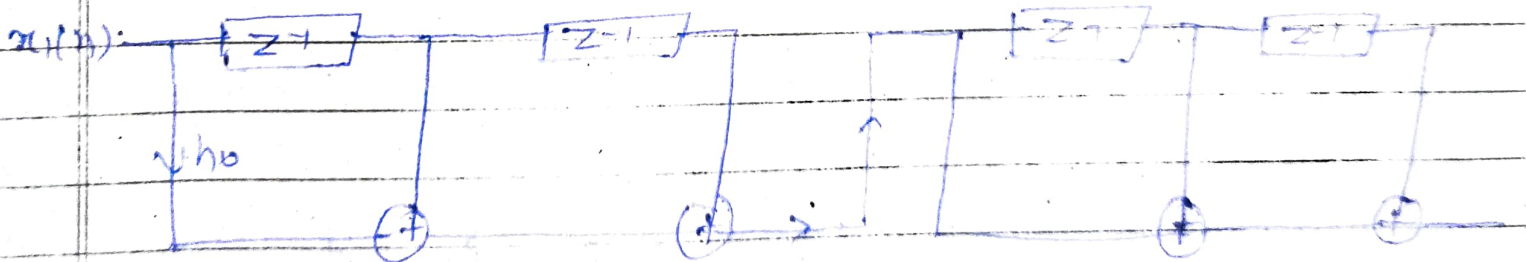
Taking inverse Z transform of both sides

$$y_K(n) = h(K_0) x_K(n) + h(K_1) x(n-1) + h(K_2) x(n-2)$$

$$y_K(n) = h(K_0) x_K(n) + h(K_1) x(n-1) + h(K_2) x(n-2) \quad \text{--- (9)}$$



$$x(n) \rightarrow x_1(n) \xrightarrow{h_1(z)} y_1(n) = x_2(n) \xrightarrow{h_2(z)} y_2(n) = x_3(n) \xrightarrow{h_3(z)} y_K(n)$$



Reduce cascade realization of

$$H(z) = \left(1 + \frac{1}{4}z^{-1} + z^{-2}\right) \left(1 + \frac{1}{8}z^{-1} + z^{-2}\right)$$

For cascade realization

$$H(z) = H_1(z) \cdot H_2(z)$$

$$H_1(z) = 1 + \frac{1}{4}z^{-1} + z^{-2} \quad H_2(z) = 1 + \frac{1}{8}z^{-1} + z^{-2}$$

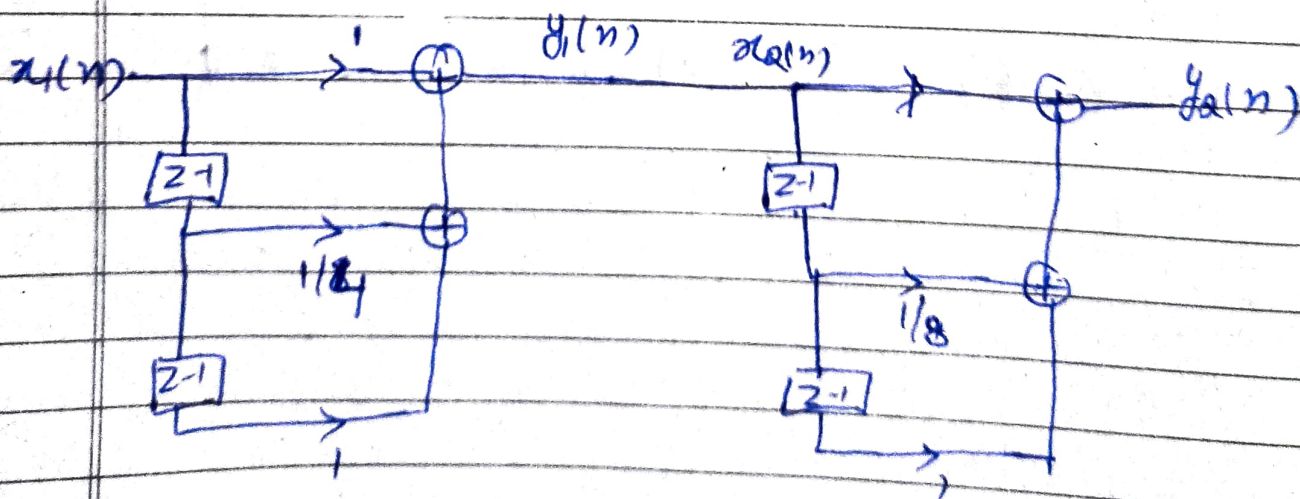
$$H_1(z) = \frac{Y_1(z)}{X_1(z)} = 1 + \frac{1}{4}z^{-1} + z^{-2}$$

$$Y_1(z) = X_1(z) + \frac{1}{4}z^{-1}X_1(z) + z^{-2}X_1(z)$$

Taking IZT

$$y_1(n) = x_1(n) + \frac{1}{4}x_1(n-1) + x_1(n-2)$$

$$y_2(n) = x_2(n) + \frac{1}{8}x_2(n-1) + x_2(n-2)$$



Obtain the system for $H(z)$ and difference eqⁿ for $h(n) = \{1, -2, -2, 3\}$ draw direct form FIR filter structure

Here range of n is $n=0$ to 3

$$H(z) = \sum_{n=0}^M h(n) z^{-n}$$

$$H(z) = h(0) + h(1)z^{-1} + h(2)z^{-2} + h(3)z^{-3}$$

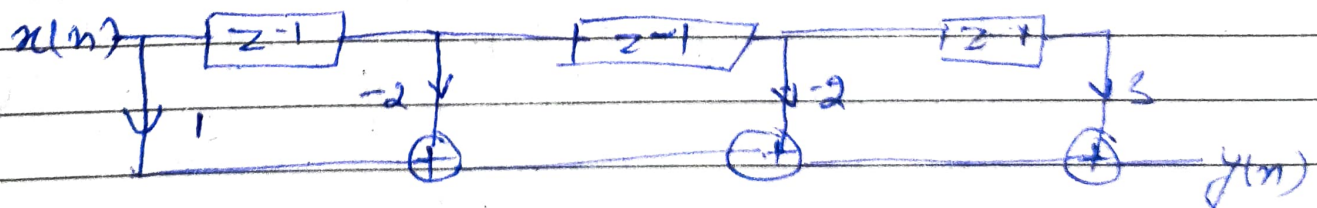
$$H(z) = 1 - 2z^{-1} - 2z^{-2} + 3z^{-3}$$

$$H(z) = \frac{Y(z)}{X(z)}$$

$$Y(z) = X(z) - 2z^{-1}X(z) - 2z^{-2}X(z) + 3z^{-3}X(z)$$

Taking IZT

$$y(n] = x(n] - 2x(n-1] - 2x(n-2] + 3x(n-3]$$



Structures for linear phase FIR filters →

FIR filter is said to be having a linear phase structure if its unit impulse sequence is either symmetric or antisymmetric about some point in time. That means FIR filter has linear phase if it satisfies the conditions

$$h(n) = h(M-1-n) \quad n=0, 1, \dots, M-1 \quad \text{--- (1)}$$

Here M = Number of samples

The transfer fn of FIR filter is given by

$$H(z) = \sum_{n=0}^{M-1} h(n) z^{-n} \quad \text{--- (2)}$$

Let us split the summation into two parts

$$H(z) = \sum_{n=0}^{\frac{M}{2}-1} h(n) z^{-n} + \sum_{n=\frac{M}{2}}^{M-1} h(n) z^{-n} \quad \text{--- (3)}$$

But for linear phase we have.

$$h(n) = h(M-1-n) \Rightarrow n = M-1-n$$

Put in ^{second} eqn (3) we get

$$H(Z) = \sum_{n=0}^{\frac{M}{2}-1} h(n) Z^n + \sum_{n=0}^{\frac{M}{2}-1} h(M-1-n) Z^{-(M-1-n)} \quad (4)$$

NOTE

The limit of Σ are changes as follows
when $n = M-1-n$

same limit $n = \frac{M}{2} \Rightarrow M-1-n = \frac{M}{2}$

$$\frac{M}{2} - M + 1 = n \quad M-1-\frac{M}{2} = n$$

$$\frac{M}{2} - 1 = n$$

when $n = M-1 = M-1-n = M-1$

$$M-1-M+1 = n$$

$$n=0$$

$$H(Z) = \sum_{n=0}^{\frac{M}{2}-1} h(n) [Z^{-n} \cdot Z^{-(M-1-n)}]$$

(3)

Case 1 For even M

We know that $H(Z) = \frac{Y(Z)}{X(Z)}$

$$\frac{Y(Z)}{X(Z)} = \sum_{n=0}^{\frac{M}{2}-1} h(n) [Z^{-n} + Z^{-(M-1-n)}]$$

$$Y(Z) = \sum_{n=0}^{\frac{M}{2}-1} h(n) [Z^{-n} + Z^{-(M-1-n)}] X(Z)$$

$$Y(z) = h(0) [1 + z^{-(M-1)}] X(z) + h(1) [z^{-1} + z^{-(M-2)}] X(z) + \dots + h\left(\frac{M}{2} - 1\right) [z^{-(\frac{M}{2}-1)} + z^{-\frac{M}{2}}] X(z)$$

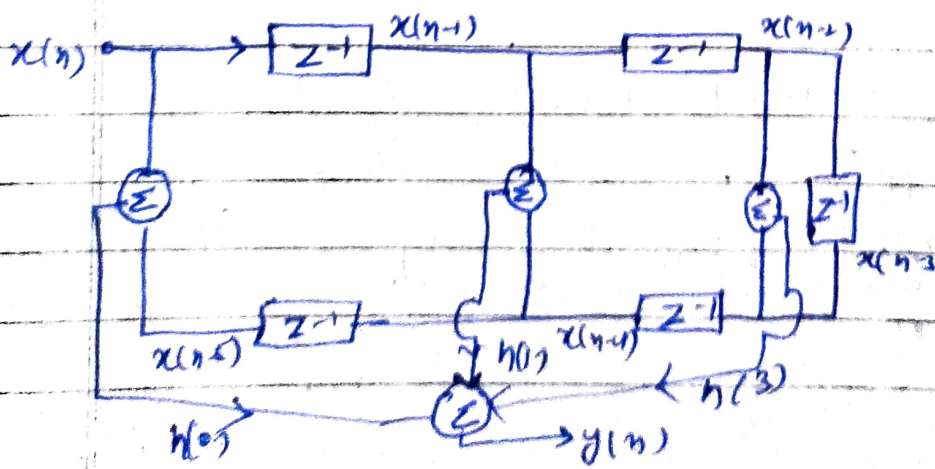
$$Y(z) = h(0) [X(z) + z^{-(M-1)} X(z)] + h(1) [z^{-1} X(z) + z^{-(M-2)} X(z)] + \dots + h\left(\frac{M-1}{2}\right) [z^{-\frac{M-1}{2}} X(z) + z^{-M/2} X(z)]$$

Taking 12T both sides

$$y(n) = h(0) [x(n) + x(n-(M-1))] + h(1) [x(n-1) + x(n-(M-2))] + \dots + h(\frac{M-1}{2}) [x(n-(\frac{M-1}{2})) + x(n-\frac{M}{2})]$$

Let $M=6$

$$g(n) = h(0) [x(n) + x(n-5)] + h(1) [x(n-1) + x(n-4)] + h(2) [x(n-2) + x(n-3)]$$



Case 2 For odd M

When M is odd let $M=5$

$$y(n) = h(0)x(n) + h(1)x(n-1) + h(2)x(n-2) + h(3)x(n-3) + h(4)x(n-4)$$

But we have the condition for symmetry
 $h(n) = h(M-1-n)$

$$M=5$$

$$n=0 \Rightarrow h(0) = h(4)$$

$$n=1 \Rightarrow h(1) = h(3)$$

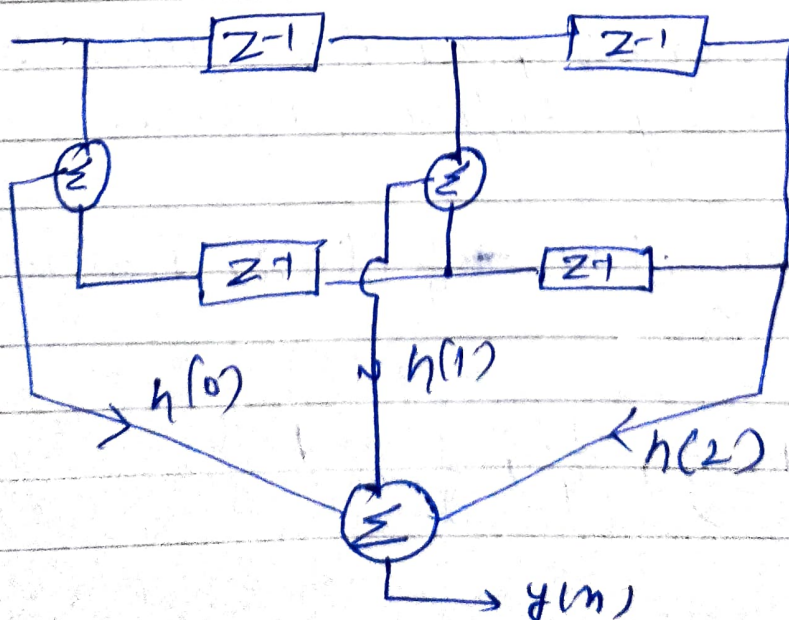
$$n=2 \Rightarrow h(2) = h(2)$$

$$n=3 \Rightarrow h(3) = h(1)$$

$$n=4 \Rightarrow h(4) = h(0)$$

$$y(n) = h(0)x(n) + h(1)x(n-1) + h(2)x(n-2) + h(3)x(n-3) + h(4)x(n-4)$$

$$y(n) = h(0)[x(n) + x(n-4)] + h(1)[x(n-1) + x(n-3)] + h(2)[x(n-2)]$$



obtain linear Phase realization of $H(z)$

$$H(z) = 1 + \frac{z^{-1}}{4} + \frac{z^{-2}}{4} + z^{-3}$$

$$h(0) = 1, \quad h(1) = \frac{1}{4}, \quad h(2) = \frac{1}{4}, \quad h(3) = 1$$

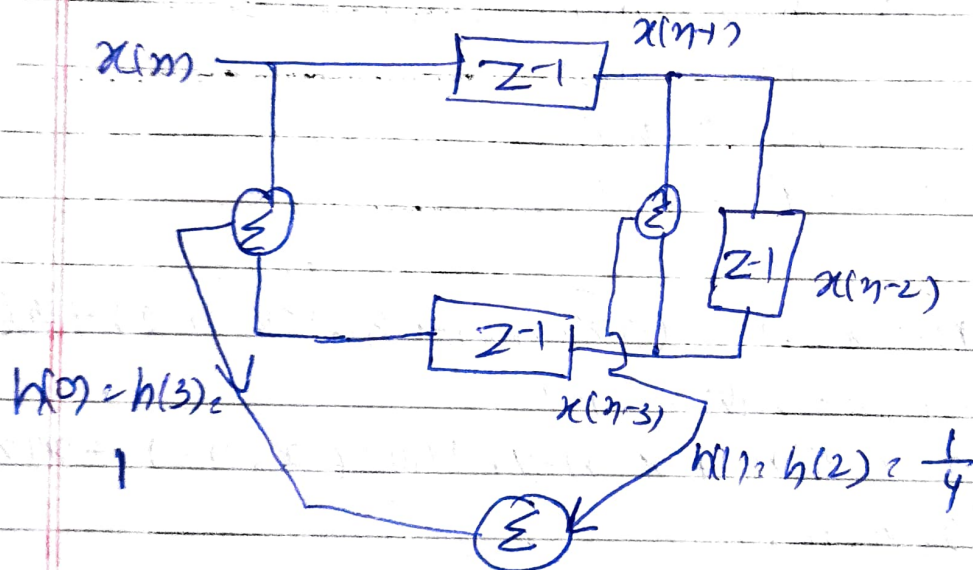
$M=4$ that means even

$$h(n) = h(M-1-n)$$

$$h(0) = h(3)$$

$$h(1) = h(2)$$

$$h(2) = h(1)$$



Realize a linear phase FIR filter with the following impulse response.

$$h(n) = \delta(n) + \frac{1}{2} \delta(n-1) + \frac{1}{4} \delta(n-2) + \delta(n-4) + \frac{1}{2} \delta(n-3)$$

$$h(n) = \left\{ 1, \frac{1}{2}, -\frac{1}{4}, \frac{1}{2}, 1 \right\}$$

Obtain linear Phase approximation
 $h(0) = 1$, $h(1) = \frac{1}{2}$, $h(2) = \frac{1}{4}$, $h(3) = \frac{1}{2}$
 $h(4) = 1$

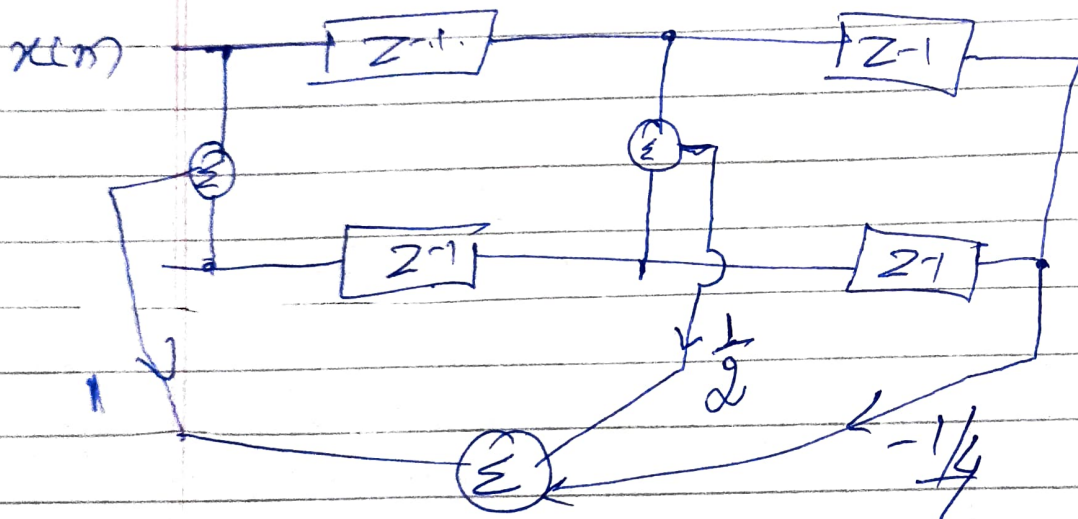
Here $M = 5$

$$h(n) = h(M-1-n)$$

$$n=0 \Rightarrow h(0) = h(4)$$

$$n=1 \Rightarrow h(1) = h(3)$$

$$n=2 \Rightarrow h(2) = h(2)$$



Realize the following system
 in using minimum no of multipliers

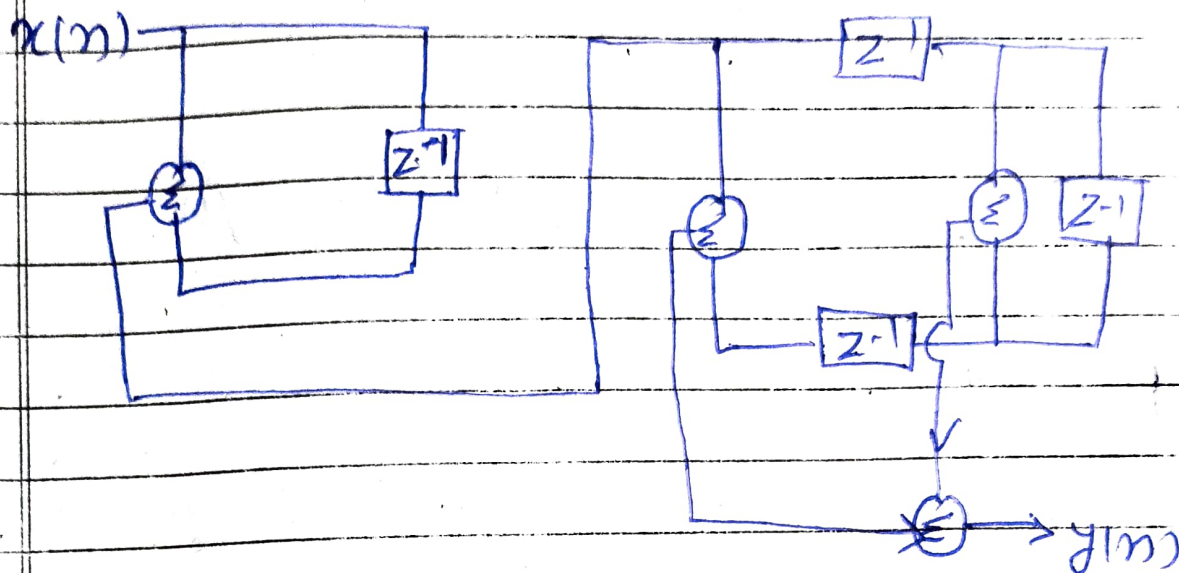
$$H(z) = (1 + z^{-1}) \left(1 + \frac{1}{2} z^{-1} + \frac{1}{2} z^{-2} + z^{-3} \right)$$

$$H_1(z) = 1 + z^{-1}$$

$$H_2(z) = 1 + \frac{1}{2} z^{-1} + \frac{1}{2} z^{-2} + z^{-3}$$

$$H_1(z) \Rightarrow h(0) = 1, h(1) = 1$$

$$H_2(z) \Rightarrow h(0) = 1, h(1) = \frac{1}{2}, h(2) = \frac{1}{2} = h(3) = 1$$



Deduce cascade realization of

$$H(z) = \left(1 + \frac{1}{4}z^{-1} + z^{-2}\right) \left(1 + \frac{1}{8}z^{-1} + z^{-2}\right)$$

For cascade realization

$$H(z) = H_1(z) \cdot H_2(z)$$

$$H_1(z) = \frac{Y_1(z)}{X_1(z)} = 1 + \frac{1}{4}z^{-1} + z^{-2}$$

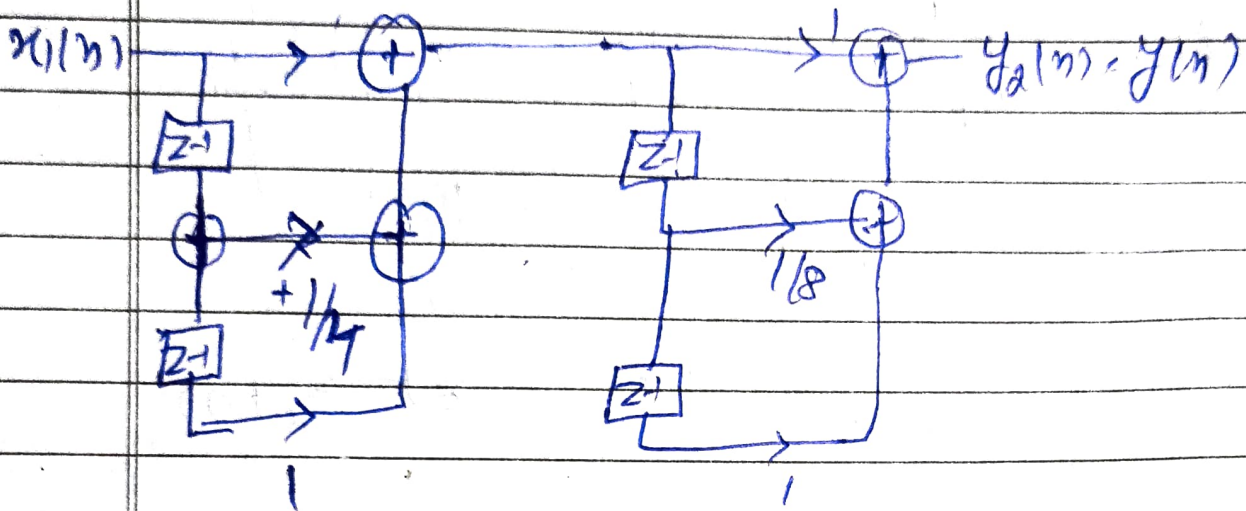
$$Y_1(z) = X_1(z) + \frac{1}{4}z^{-1}X_1(z) + z^{-2}X_1(z)$$

$$Y_2(z) = X_2(z) + \frac{1}{8}z^{-1}X_1(z) + z^{-2}X_2(z)$$

Taking 1 ZT we get

$$y_1(n) = x_1(n) + \frac{1}{4} x_1(n-1) + x_1(n-2)$$

$$y_2(n) = x_2(n) + \frac{1}{8} x_2(n-1) + x_2(n-2)$$



Obtain the system in $H(z)$ and difference eqn for $h(n) = \{1, -2, -2, 3\}$
draw direct form FIR filter structure

$$H(z) = \{1, -2, -2, 3\}$$

range of n is $n = 0$ to 3

using Z transform

$$H(z) = \sum_{n=0}^M h(n) z^{-n}$$

$$h(0) + h(1)z^{-1} + h(2)z^{-2} + h(3)z^{-3}$$

$$H(z) = 1 + 2z^{-1} - 2z^{-2} + 3z^{-3}$$

Calculations of difference eqn.

$$H(z) = \frac{Y(z)}{X(z)}$$

$$Y(z) = X(z) - 2z^{-1}X(z) - 2z^{-2}X(z) + 3z^{-3}X(z)$$

taking IZT

$$y(n] = x[n] - 2x[n-1] - 2x[n-2] + 3x[n-3]$$

